Ultrafast Phenomena: A Laboratory Experiment For Undergraduates

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Ultrafast phenomena: A laboratory experiment for undergraduates

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We present a set of experiments that introduce students to ultrafast science. We discuss the relationship between the description of an ultrashort laser pulse in the frequency domain and the time domain. Using experimental results we demonstrate that this relationship is constrained by the lower limit of the Heisenberg uncertainty principle. Students carrying out the experiments will become familiar with ultrafast techniques, such as autocorrelation and laser cavity design, as well as various other concepts such as dispersion, Fourier transformation, interference, and nonlinear optics.

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I. INTRODUCTION

“Ultrafast” is a term typically used to describe processes which occur on a picosecond ($10^{-12}$ s) or faster time scale. Up to about 1965 the shortest time interval that could be resolved was 1 ns ($10^{-9}$ s), a limit set by the capabilities of high frequency electronic circuitry. A century earlier, time resolution of only milliseconds was available from streak-recording methods. The time resolution limit dropped precipitously after 1965 due to the invention of the laser and now is on the order of 10 fs ($10^{-14}$ s).

The techniques and technology of ultrafast science provide a unique method for investigating physical phenomena throughout the physical and biological sciences. For example, in 1991 Schoenlein et al. measured a 200-fs time constant for the 11-cis to 11-trans torsional isomerization of the rhodopsin chromophore, the first step in the vision process. Investigations of photosynthesis showed the lifetime of the excited chlorophyll b molecule to be 15 ps at concentrations near those of living cells (0.1 M); these measurements can be used to infer the structure of the chlorophyll b molecule. A number of DNA studies have also been carried out, some, for instance, studying the transfer of energetic excitations along the DNA molecule. Recently, the first study of protein dynamics using picosecond infrared vibrational echo experiments was carried out by Rella et al. They studied myoglobin-CO (used in storage and transport of O$_2$ in muscle tissue) and showed the existence of many confor-
II. THEORY

The Heisenberg uncertainty principle (HUP) states that there is a limit to the information we can acquire about conjugate variables of a system. For example, our simultaneous knowledge of both the position, \(x\), and momentum, \(p\), of a system is limited:

\[
\Delta x \Delta p \geq \hbar/2.
\]

Here, “\(\Delta\)” indicates the “uncertainty” in the value of the quantity being measured. Mathematically,

\[
\Delta x = \sqrt{\langle x^2 \rangle - \langle x \rangle^2},
\]

where the brackets define an expectation value which is given by

\[
\langle x \rangle = \langle \Psi | x | \Psi \rangle = \int_{-\infty}^{\infty} \Psi^* x \Psi \, dx.
\]

A similar relation holds for the energy, \(E\), and time, \(t\), variables of a system:

\[
\Delta E \Delta t \geq \hbar/2.
\]

For photons \(\Delta E = h \Delta \omega\), so Eq. (4) can be rewritten as

\[
\Delta \omega \Delta t \geq 1/2.
\]

In the context of a laser pulse, \(\Delta \omega\) refers to the spectral bandwidth of the pulse and \(\Delta t\) describes the pulse’s temporal duration. It should be recalled that implicit in the word “pulse” is the idea of localization. For a pulse (i.e., a statistically large group of photons) to be localized in time it must contain many frequency components, as is known from Fourier decomposition, or, more qualitatively, from Eq. (5). In fact, the localization of a group of photons in an interval of time necessarily sets a limit to the minimum probable range of photon energies. The theoretical basis for this relation derives from the most general form of the Heisenberg uncertainty principle

\[
\Delta A^2 \Delta B^2 \geq \frac{1}{4} \left[ \langle [\hat{A}, \hat{B}] \rangle \right]^2.
\]

The time evolution of an expectation value is given by

\[
\frac{d}{dt} \langle A \rangle = \frac{i}{\hbar} \left[ H, \langle A \rangle \right] = \left( \frac{\partial}{\partial t} \right) \langle A \rangle.
\]

The position–momentum uncertainty relation follows directly from Eq. (6). However, deriving the energy–time uncertainty relation is not as straightforward since \(\hat{t}\) is not a valid operator. Even so, an expression analogous to \(\Delta t \Delta \omega \geq \hbar/2\) can be developed for energy and time as follows. For an operator that does not depend explicitly on time \([\langle \partial / \partial t \rangle \hat{A} = 0]\) we can rewrite the general form of the uncertainty relation. Taking \(\hat{B} = \hat{H}\), and using Eq. (7), we can re-write Eq. (6) as

\[
\Delta E \left( \frac{\Delta A}{d/dt \langle A \rangle} \right) \geq \frac{\hbar}{2}.
\]

The term in the brackets can be identified as time uncertainty, \(\Delta t\), if we interpret \(\Delta t\) as a measure of the time it takes for the expectation value of observable \(A\) to change significantly. For example, \(\Delta t\) for a laser pulse with a Gaussian spatial profile may be considered as the amount of time it takes for the position of the pulse to change by the spatial width of the pulse, \(\sigma_x\). Thus we arrive at the energy–time uncertainty relation in Eq. (4) taking

\[
\left( \frac{\Delta A}{d/dt \langle A \rangle} \right) = \Delta t.
\]

Looking at Eq. (5) it is clear that a laser pulse of shorter temporal duration can be created if more spectral bandwidth is added to the pulse. To generate short pulses, therefore, we need a lasing medium which can supply a large gain bandwidth and a laser cavity which can support that bandwidth. The most widely used medium for supplying the bandwidth is titanium-doped sapphire (Ti:Al̂2O₃) which lases over the range of approximately 700–1000 nm. This material can
provide gain to over $10^6$ longitudinal cavity modes for a
typical laser cavity. (The $n$th longitudinal cavity mode has
wavelength $2L/n$, where $L$ is the length of the laser cavity
and $n$ is an integer.) The $i$th mode can be described mathemat-
ically as $A \sin(n \pi z/L + \phi_i)$, where $A_i$ is the amplitude
and $\phi_i$ is the phase at $z=0$). This can be compared to a
helium–neon laser which has a gain bandwidth on the order
of $10^{-2}$ nm and therefore lases in only a few longitudinal
cavity modes.

A large spectral bandwidth, however, is not a sufficient
condition for ultrashort pulse generation. Each spectral com-
ponent (longitudinal mode) must have a specific phase rela-
tion to the other components, i.e., the pulse must be “mode
locked.” This is illustrated in Fig. 1, a simulation of the
pulse intensity resulting from the addition of 20 cavity
modes. In one case, a random phase is assigned to each mode
before the modes are added together and the result is a rela-
tively flat intensity versus time. In fact, the intensity versus
time tends toward a constant value as the number of random-
phase cavity modes is increased. Alternatively, when a par-
ticipate phase relation is set among the modes
—e.g., the phase
$\phi_i$ of the $i$th mode is zero for each wave in this case, although any constant
phase for each mode, or a random phase relationship. Note that a pulse is
said to be “mode locked,” as each of the modes has a spe-
cific phase relationship to one another.

In practice, mode locking the Ti:Al$_2$O$_3$ pulse is achieved
through Kerr-lens mode locking, a process which exploits
the optical Kerr effect. Kerr-lens mode locking is utilized
in a laser cavity designed to have higher gain for a pulse that
has a short duration versus one of longer duration. This is
possible because a short pulse will have a higher intercavity
intensity than a long pulse, and pulses of high enough intensity
can undergo nonlinear interactions with the Ti:Al$_2$O$_3$.
This nonlinear interaction is exploited to produce high gain
per roundtrip in the cavity.

For a further discussion of cavity modes, gain bandwidth,
mode locking, and lasers in general, we recommend Lasers
by A. Siegman$^{23}$ and Lasers by P. Milonni and J. Eberly. We
now discuss some of the properties of ultrashort pulses,
particularly the relationship between the time domain
and frequency domain descriptions of a pulse. In order to
simplify our quantitative discussion, we first consider pulses
with Gaussian envelopes. The Gaussian shape is convenient
because the Fourier transform of a Gaussian function is an-
other Gaussian function. It is important to note that the
discussion we present can be easily generalized to any pulse
shape, as shown below in the analysis section.

The electric field of a Gaussian pulse is given by

$$E(t) = E_0 \exp \left(\frac{-i^2}{2 \sigma_t^2}\right) \exp(-i \omega_0 t),$$

where $E_0$ is real, $\omega_0$ is the average, or “carrier,” frequency
of the wave (near $10^{15}$ Hz for a Ti:Al$_2$O$_3$ laser), and $\sigma_t$ is a
measure of the pulse width in time. The full width at half-
maximum (FWHM) of the pulse duration is

$$t_{\text{FWHM}} = 2 \sqrt{\ln(2)} \sigma_t.$$  

Note that this is the FWHM of the intensity of the pulse
($\approx E^2$), not of the field.

We can write an equation completely analogous to Eq. (9)
to describe the pulse in the frequency domain:

$$\mathcal{F}(\omega) = \mathcal{F}_0 \exp \left(-\frac{(\omega - \omega_0)^2}{2 \sigma_{\omega}^2}\right).$$

The FWHM of the intensity of the pulse’s spectral width is
given by

$$\omega_{\text{FWHM}} = 2 \sqrt{\ln(2)} \sigma_{\omega}. $$

From Eq. (2) we can calculate the “uncertainty” of our pulse
in time and frequency domains. We identify the “uncer-
tainty” as a measure of the pulse width in the particular
domain. Using an equation analogous to Eq. (3) to calculate
$\langle \omega \rangle$ and $\langle \omega^2 \rangle$, we find

$$\langle \omega^2 \rangle = \omega_0^2 + \frac{\sigma_{\omega}^2}{2}, \quad \langle \omega \rangle = \omega_0,$$

so that

$$\Delta \omega = \frac{\sigma_{\omega}}{\sqrt{2}}.$$ 

An identical expression holds for $\Delta t$. Thus Eq. (5) becomes

$$\Delta \omega \Delta t = \frac{\sigma_{\omega} \sigma_{\tau}}{2} \geq \frac{1}{2},$$

or

$$\sigma_{\omega} \sigma_{\tau} \geq 1.$$ 

Thus $1$ is the minimum time-bandwidth product of a Gauss-
ian shaped pulse. It’s important to realize that the minimum
value given here depends on (i) the shape of the pulse enve-
lopes being Gaussian and (ii) the definition of $\Delta \omega$ and $\Delta t$
given above (as compared to, say, FWHM). A Gaussian
pulse which meets the minimum time-bandwidth constraint
is said to be “time-bandwidth limited,” or “transform lim-
ited.” In the case of ultrashort pulses, we hope to generate
transform limited pulses, since we want the minimum pulse
duration for a given spectral bandwidth.

This does, however, bring up the question: “Under what
conditions won’t a pulse be transform limited?” To explore
the answer to this question, consider the oscillation of a spec-
tral component, $\omega$, of the pulse: $\cos(k_w z - \omega t + \phi(\omega))$, where
$k_w$ is the wave number of the $\omega$ component. $\phi(\omega)$ is the
component’s phase which can be expanded as a Taylor series
about the carrier frequency:
\[ \phi(\omega) = \phi(\omega_0) + \left( d\phi \right)_{\omega_0} (\omega - \omega_0) \\
+ \frac{1}{2!} \left( \frac{d^2\phi}{d\omega^2} \right)_{\omega_0} (\omega - \omega_0)^2 + \cdots. \tag{17} \]

(Please see the Appendix for the derivation and a physical discussion of this expression.) Pulses are transform limited if the coefficients of the quadratic and higher order terms are zero.

In general this won’t be the case and the result is that, in the time domain, the short pulse will have a time dependent frequency. We are therefore motivated to more carefully consider Eq. (9). The harmonic term, \( \exp(-i\omega f) \), in a limited validity as it stands because it describes only a single frequency pulse—a pulse which if described by Eq. (17) would have all terms of second order or higher equal to zero. Equation (9) can be made to describe a pulse with a time dependent frequency if we construct a more general expression for the argument of the exponent. We replace \( \omega_0 t \) with \( g(t) \), where

\[ g(t) = g(t_0) + \left( \frac{dg}{dt} \right)_{t_0} (t-t_0) + \frac{1}{2!} \left( \frac{d^2g}{dt^2} \right)_{t_0} (t-t_0)^2 + \cdots. \]

It is possible to determine \( g(t) \) for a given pulse, and from this the instantaneous frequency of the pulse as a function of time, since

\[ \omega(t) = \frac{d}{dt} g(t). \tag{19} \]

If all the terms beyond the linear terms in Eqs. (17) and (18) are zero, then the pulse is described by a single carrier frequency \( \omega_0 \) [see Eq. (19)]. If any terms beyond the linear term are nonzero, then the pulse has a time dependent frequency and does not conform to the minimum constraint of the HUP.

In particular, if the quadratic term of the Taylor series is nonzero, the pulse is linearly dispersed, or “chirped,” since the instantaneous frequency is linear in time [see Eq. (19)]. Accumulation of the quadratic term is physically due to the dispersion of the optical components (prisms, mirrors, lenses, beam splitters, etc.) with which the pulse interacts as it propagates. Control of this parameter, in both magnitude and sign, is a critical part of the technical operation of ultrashort laser pulse experiments.226 The terms beyond the quadratic become relatively more important as the quadratic term is minimized and are critical to reduce in order to drop the pulse duration below 100 fs.

In summary, a Ti:Al\(_2\)O\(_3\) laser is capable of generating a pulse with spectral bandwidth \( \sigma_s \). Given this bandwidth, the minimum temporal duration of the pulse is given by the HUP. Thus, if one can generate transform limited pulses, laboratory experiments can be built which demonstrate and quantify this basic principle of quantum mechanics.

III. EXPERIMENTAL SETUP

We start this section with some words of caution. The peak of the Ti:Al\(_2\)O\(_3\) laser emission is in the near infrared, however, there is enough energy in the red that the beam can be seen by the unaided eye. Looking at the beam in this way, one might be led to believe that the laser is weak, as the visible portion of the emission is faint. This is incorrect. The laser has an average output power of 600 mW and a peak power of over \( 10^5 \) W. Significant and permanent retinal damage will occur if the beam enters the eye. Throughout all stages of the laboratory experiment, proper safety goggles should be worn by students and faculty.

The experimental layout, shown in Fig. 2, consists of an argon ion laser, Ti:Al\(_2\)O\(_3\) laser oscillator, external-prism pulse compressor, autocorrelator, and spectrometer. Central to the laboratory is the Ti:Al\(_2\)O\(_3\) laser27 pumped with an argon ion laser intensity stabilized at 4.5 W (all lines). The argon laser beam maintains a fixed propagation direction with an external pointing stabilizer.28 Maintaining spatial and pointing properties of the pump laser is critical in providing intensity and mode locking stability of the ultrashort pulse laser.

Alignment of the Ti:Al\(_2\)O\(_3\) oscillator is an important part of the laboratory, and we consider it to be the students’ first experiment. The central concern in aligning the laser cavity is to minimize phase distortion, i.e., make all the nonlinear terms in Eqs. (17) and (18) zero. To aid in this, one can manipulate the “amount of glass” in the beam, i.e., the position of the compensation prisms along an axis that bisects the prism apex and is in a plane perpendicular to the prism surfaces. The prisms provide negative dispersion and can be used to control the magnitude and sign of the dispersion compensation within the laser cavity. More glass means more positive dispersion compensation. Mode locking stability is maintained when the overall cavity (compensation prisms plus gain medium and mirrors) has near zero net dispersion. This is achieved with the prisms set to balance the natural positive dispersion of the gain medium.226 Beyond the zero-dispersion condition, the prism material must be carefully chosen to minimize the third and fourth order phase distortion. Though the oscillator has a roundtrip dispersion balance, the output has a positive chirp caused, in part, by the pulse’s passage through the output coupler. The zero-chirp condition inside the cavity is designed to occur inside the gain medium in order to best effect Kerr lensing.

Once a stable mode locked output is attained from the oscillator, its spectrum and autocorrelation traces are measured as the second and third experiments carried out by the students. Measuring the spectrum is a good introduction into the physical properties of mode locked pulses and it helps to illustrate the relations given by the HUP. Mode locked pulses, having a short temporal duration, display a very broad spectrum compared to the continuous wave spectra of a single-mode laser. Broadband measurements are done using a linear-array-detector spectrometer.30 The spectrum can be read directly from the laser output (after heavy attenuation) or from a scrap reflection in the autocorrelator. Given the pulse spectrum, a quick, back-of-the-envelope calculation can give an idea of the pulse duration since

\[ \Delta f = \frac{c}{\lambda^2} \Delta \lambda \]  

and \( \Delta t \approx 1/\Delta f \).

Following the spectral measurements students embark on the more difficult pulse duration measurement. Output from the Ti:Al\(_2\)O\(_3\) laser is filtered and sent through a double pass, external-prism-pair compressor to compensate for the chirp inherent in the laser output, from the output coupler and the IR pass filter, and from the beam splitters in the autocorrelator. Both external prisms are aligned at the Brewster angle and mounted on translation stages so they can maintain this
alignment as they are translated through the beam. This configuration allows for adjustment of chirp, which leads to observable changes in the pulse correlation time, envelope ratio, and envelope shape. It is worth noting that we have found that the tip-to-tip prism separation is greater for the external prism pair than for the intercavity pair. This is indicative of the larger degree of “prechirp” needed to compensate for dispersion in the autocorrelator optics which the pulse will encounter. All mirrors used in this experiment are silver-coated substrate with a protective magnesium-fluoride layer. Silver is the preferable coating since it has a broad and flat reflective bandwidth over the wavelengths of interest, and thus does minimal modification of the spectral content of the pulse.

The autocorrelator design shown in Fig. 2 is based on previously reported configurations.26 The input beam is split into two beams with a front surface, 50/50 beam splitter (back surface is antireflection coated). Each beam is retroreflected with aluminum mirror corner cubes and recombined with a second, separate beam splitter. The use of two beam splitters is important in that each path in the interferometer will contain the same amount of beam-splitter glass, and therefore the same amount of dispersion. Thus the external-prism pair is aligned to prechirp the pulse such that the minimum pulse width will be measured by the autocorrelator. Presumably, the minimum chirp occurs at the harmonic crystal where the correlation function is generated. The interference aspect of this technique depends critically on the alignment of the beam splitters, and each is on a separate tilt mount. A quick check of the alignment can be made by looking at the beam overlap at a distant location (several meters) with an IR sensitive card. The combined beams are focused into a 50-μm-thick second harmonic generation (SHG) KDP crystal31 with a concave, silver-coated mirror (focal length=10 cm). The SHG is collected with a second concave mirror positioned to focus the light onto a photomultiplier tube (PMT). Several blue-pass filters are placed in front of the PMT to eliminate the fundamental, IR laser light. The photocurrent is converted into a voltage with a 1000-Ω resistor and read on a digital oscilloscope.

One of the corner cubes is mounted on a fine translation stage adjusted for a match of the interferometer path lengths. The other corner cube is mounted on the cone of an audio speaker. By driving the speaker with the amplified triangle wave of a function generator the interference pattern is scanned at audio frequency repetition rates. This rate is tens of thousands times slower than the pulsing rate of the laser, so the PMT signal is approximately continuous throughout the scan. Too large a translation amplitude can reduce the interference fringe contrast, so care must be taken not to tune the audio frequency too close to a low frequency resonance of the speaker/cube system. Though audio speakers are designed for a relatively flat frequency response, the large aluminum housing of the corner cube glued to the cone will change these fine characteristics and low frequency resonances may develop. Precise position of the corner cube is
not important in this technique since the physical translation is inferred from the fringe peaks. The separation between peaks corresponds to a net path change of one wavelength of the fundamental frequency, and this wavelength is known from the spectrum.

Relating the spectrum measured in the second experiment to the autocorrelation trace measured in the third is the subject of the following section.

IV. ANALYSIS

This section describes the steps of analysis used to determine if a given laser pulse is at the limit of the HUP, i.e., whether or not it is transform limited. Starting with the power spectrum of the pulse, we take its square root to generate a spectral amplitude followed by a Fourier transform to generate the time amplitude (electric field) of the pulse. The square of the time amplitude is proportional to the intensity of the pulse, assuming that the pulse was originally transform limited. This assumption must be made because the square root of the spectrum, unfortunately, is not a complete description of the pulse in the frequency domain. It measures the intensity of each frequency component, but indicates nothing about each component’s phase. To carry out a Fourier transformation from the frequency domain to the time domain we need to know the phase information, and this is not easily obtained. To work around this, we assume the phase of each frequency component is zero; in the context of the interferometric autocorrelation, this assumption constrains the pulse to have the properties of a transform limited. This assumption must be made because the square of the time amplitude is proportional to the intensity of the pulse. The time averaged intensity of the KDP’s emission at $2\omega_0$, as defined in Eq. (9), establishes a nonlinear dipole moment, $P^{2\omega_0}$, in the crystal at frequency $2\omega_0$. The power series expansion of the dipole moment in terms of the field strength is

$$P(t) = \chi^{(1)}E(t) + \chi^{(2)}E^2(t) + \chi^{(3)}E^3(t) + \cdots$$

$$= P_0(t) + P^{1\omega_0}(t) + P^{2\omega_0}(t) + \cdots,$$

where the $\chi^{(i)}$ are the nonlinear optical susceptibility tensors. Therefore, the nonlinear dipole moment at $2\omega_0$ is proportional to the square of the sum of the fields:

$$P^{2\omega_0} \propto [E(t) + E(t - \tau)]^2.$$  

The time averaged intensity of the KDP’s emission at $2\omega_0$, $\langle I^{2\omega_0} \rangle$, is proportional to the complex square of the dipole at that frequency:

$$\langle I^{2\omega_0} \rangle \propto |P^{2\omega_0}|^2$$

$$\propto [E(t)^2 + E(t - \tau)^2 + 2E(t)E(t - \tau)] \times [E^*(t)^2 + E^*(t - \tau)^2 + 2E^*(t)E^*(t - \tau)].$$

Integrating this result from $t = -\infty$ to $t = +\infty$, for the Gaussian pulse of Eq. (9), we arrive at

$$\langle I^{2\omega_0}(\tau) \rangle \propto \left[ 1 + \exp \left( \frac{-\tau^2}{2\sigma_r^2} \right) \right] \left[ 2 + \cos(2\omega_0\tau) \right]$$

$$+ 4 \exp \left( \frac{-3\tau^2}{8\sigma_r^2} \right) \cos(\omega_0\tau).$$

This is the interferometric autocorrelation signal measured at the photomultiplier tube in Fig. 2.

A similar result can be derived for Gaussian pulses which have a chirp [$\alpha \neq 0$ in Eq. (18)]. In this case,

$$E(t) = E_0 \exp \left( -\frac{t^2}{2\sigma_i^2} \right) \exp[i(\omega_0 + \alpha t)t].$$

By the same process we went through to arrive at Eq. (23), we find

$$\langle I^{2\omega_0}(\tau) \rangle \propto \left[ 1 + \exp \left( \frac{-\tau^2}{2\sigma_r^2} \right) \right] \left[ 2 + \cos(2\omega_0\tau) \right]$$

$$\times \exp(-2(\sigma\alpha\tau)^2) \times \exp \left( \frac{-3\tau^2}{8\sigma_r^2} \right)$$

$$\times \exp \left( -\frac{(\alpha\tau)^2}{2} \right) \cos(\omega_0\tau) \cos(\frac{\alpha^2\tau^2}{2}).$$

We can work out the expression for $\langle I^{2\omega_0} \rangle$ numerically for any pulse shape and with any order of phase distortion in time. This is done by defining an electric field

$$E(t) = A(t) \exp(i[\omega_0 + \alpha t + \beta t^2 + \cdots]t)$$

follows this algorithm was used to generate the autocorrelation signal given the pulse spectrum as input and can be obtained directly from Tom Donnelly or by downloading from Carl Grossman’s website, http://laser.swarthmore.edu/html/research/grossman/index.html.

Below we outline the theory relevant to the interferometric autocorrelation. As shown in Fig. 2, the pulse is split in the autocorrelator and the two resulting pulses pass collinearly through the KDP crystal. The pulses may be delayed with respect to one another so we represent their electric fields as $E(\omega_0,t)$ and $E(\omega_0,t-\tau)$, as defined in Eq. (9). These fields establish a nonlinear dipole moment, $P^{2\omega_0}$, in the crystal at frequency $2\omega_0$. The power series expansion of the dipole moment in terms of the field strength is

$$P(t) = \chi^{(1)}E(t) + \chi^{(2)}E^2(t) + \chi^{(3)}E^3(t) + \cdots$$

$$= P_0(t) + P^{1\omega_0}(t) + P^{2\omega_0}(t) + \cdots,$$
and a similar electric field delayed by a time \( \tau \), \( E(\tau - \tau) \), where \( A(\tau) \) is the appropriate envelope function. The time averaged intensity of the KDP emission is then calculated numerically as described by Eqs. (21) and (22): We fix one field in time \( (\tau = 0) \) and then we calculate the square of the nonlinear dipole moment at every \( \tau \) \( (\tau = \tau + \infty) \), giving \( \langle J^2 \rangle \). The algorithm for this numeric calculation can be used to calculate the appropriate autocorrelation signal for any spectrum that is initially measured. There is no longer a need to assume that the pulse is Gaussian or otherwise. This analysis consists of three basic procedures that can be executed in most computer data analysis programs. (1) A data transformation procedure prepares the raw data, an array of spectral intensities and corresponding frequencies, into a field amplitude with appropriate array dimensions and frequency interval for a fast Fourier transform (FFT). (2) FFT converts the field amplitude spectrum to the complex, transform limited pulse envelope of the field amplitude. The magnitude of this complex envelope is used in the third procedure. (3) A simulation of the autocorrelation data is made by calculating the square of Eq. (21), the nonlinear optical polarization, for a set of interferometric time delays.

V. RESULTS AND DISCUSSION

Typical data for the experiments described above are shown in Figs. 3 and 4. The spectrum of a short pulse is shown in Fig. 3(a) and the implied transform limited pulse intensity is shown in Fig. 3(b). We have fit the spectrum to a Gaussian envelope and calculated the corresponding transform limited pulse intensity for comparison since this will represent the pulse of minimum possible duration. The autocorrelation traces that are generated from the pulses in Fig. 3(b) are shown in Fig. 4. Here we compare the measured autocorrelation trace [corresponding to the measured spectrum of Fig. 3(a)] with that which is calculated numerically from both the measured spectrum and the Gaussian fit to the spectrum. We assign a time axis to the measured autocorrelation trace by noting that the interference fringes are separated by one period of oscillation. The period of oscillation can be determined from the peak of the spectrum, e.g., Fig. 3(a) shows the spectral peak at \( 3.68 \times 10^{13} \) Hz, corresponding to a period of 2.72 fs.

It is evident that the autocorrelation trace numerically generated from the measured spectrum is slightly broader than that which is actually measured. This results from the limitation of our spectrum analyzer. The response of the detector in the spectrum analyzer is not flat as a function of frequency and thus distorts the actual spectrum. Further, the detector resolution is such that we are unable to generate a signal to noise ratio of greater than 100; this limits our ability to gain information about the spectral wings. To overcome this, we assume that the spectral intensity is zero beyond what we can reliably measure in the spectral wings. Further, since the time resolution of the fast Fourier transform in the numeric analysis is equal to the inverse of the total spectral frequency, a time resolution on the order of 1 fs requires a bandwidth \( > 10^{14} \) Hz. To achieve this bandwidth, we truncate the spectrum as described above and then add zeros to the measured spectrum below approximately \( 3.4 \times 10^{14} \) Hz and above approximately \( 4.1 \times 10^{14} \) Hz. These problems with the spectrum will be avoided in the future by taking high signal to noise measurements well into the spectral wings using a spectrometer and calibrated PMT. The distortion of the spectrum we input into the numeric analysis program accounts for the generated autocorrelation trace being slightly broader than the measured one.

The curves of Figs. 3(a) and 4(b) show that if a Gaussian shape is assumed for the spectrum the resulting numeric autocorrelation trace is nearly identical to the measured trace [see Fig. 4(b)]. This is evidence that, although our spectral
Thus from the measured autocorrelation trace in Fig. 4, smaller FWHM. In fact, trace~ in Fig. 5 are evident: As the time delay between the correlating pulses increases, the distorting effects of chirp on the autocorrelation trace~ are at the minimum constraint of the HUP.

It should be pointed out that the autocorrelation trace of a transform limited pulse is easily identified. Figure 5 compares the autocorrelation traces [see Eq. (25)] that would be generated by a Gaussian pulse with varying degrees of chirp. The transform limited ($\alpha = 0$) autocorrelation trace is shown in Fig. 5(a) and is identical to the trace of Fig. 4(b). We then add increasing amounts of chirp up through Fig. 5(f); in all cases third order and higher phase perturbations are zero. The distorting effects of chirp on the autocorrelation trace are evident: As the time delay between the correlating pulses is swept, different frequencies in each pulse interfere and thus the single frequency interference pattern of trace (a) is wiped out. Autocorrelation traces corresponding to nontransform limited pulses can be identified because they lack oscillatory behavior in the trace’s wings. A casual observer may identify trace (f) as corresponding to a shorter duration pulse than trace (a) since, at a glance, it appears to have smaller FWHM. In fact, trace (a) corresponds to the shorter, transform limited pulse.

APPENDIX

Consider a pulse at the plane $z = 0$. The field of the pulse can be approximated as:

$$E(0,t) = A(t)\exp[i\psi(t)]\exp[-i\omega_0 t],$$

(A1)

where $A$ is the real amplitude and $\psi$ is the phase modulation term. The approximation is valid as long as the range of the frequency component which make up the pulse is small compared to $\omega_0$. We let the pulse propagate through an arbitrary system and calculate the description of the output pulse at some time $t'$. This is done by taking the Fourier transform of Eq. (A1), multiplying by $\exp[i\phi(\omega)]$, where $\phi(\omega)$ is the phase shift through the system, and then taking the inverse Fourier transform:

$$E_{\text{out}}(t') = \int \frac{d\omega}{2\pi} \exp[-i\omega t']\exp[i\phi(\omega)]$$

$$\times \int dt \exp[i\omega t]A(t)\exp[i\psi(t)]\exp[-i\omega_0 t].$$

(A2)

Knowing the form of the input pulse, we can calculate the description of the output pulse if we know $\phi(\omega)$.

Physically, $\phi(\omega)$ accounts for the phase accumulated, relative to the carrier frequency, by each frequency component as the pulse propagates through the system:

$$\phi(\omega) = (k_w - k_{w_0})z.$$  

(A3)

Thus $\phi(\omega) = 0$ if the pulse propagates through vacuum. Given that the propagation constants are expressed as $k_w = n_w \omega/c$ and that the refractive index, $n_w$, can be expanded in a Taylor series about $\omega_0$, the phase term is expressed as

$$\phi(\omega) = \frac{z}{c}[\omega n_{w_0} - \omega_0 n_{w_0}]$$

$$= \frac{z}{c}\left[\omega \left(n_{w_0} + \frac{dn_w}{d\omega}\right)_{w_0} (\omega - \omega_0) + \frac{1}{2} \frac{d^2n_w}{d\omega^2}\left|_{w_0}\right.\right]$$

$$\times (\omega - \omega_0)^2 + \cdots - \omega_0 n_{w_0}.$$  

(A4)

If the index of refraction is accurately known as a function of frequency for the materials through which the beam will propagate, then Eq. (A4) allows the calculation of the coefficient of the expansion of $\phi(\omega)$ shown in Eq. (17). The physical significance of each expansion coefficient in Eq.

VI. CONCLUSION

We have presented experiments which introduce students to some of the techniques and technology used in ultrashort science. The experiments demonstrate a connection between ultrashort laser pulses and quantum mechanics, and introduce students to numeric analysis and optics. We have worked out the mathematical details for the analysis of the experiments, and have included a numeric model which may be of use. We expect these experiments to be useful in training students for future endeavors, both academic and technology related.
The permanence of mathematical achievement

What we do may be small, but it has a certain character of permanence; and to have produced anything of the slightest permanent interest, whether it be a copy of verses or a geometrical theorem, is to have done something utterly beyond the powers of the vast majority of men.